



Ultrasound Variable-Gain Amplifier

MAX2035

General Description

The MAX2035 8-channel variable-gain amplifier (VGA) is designed for high linearity, high dynamic range, and low-noise performance targeting ultrasound imaging and Doppler applications. Each amplifier features differential inputs and outputs and a total gain range of typically 50dB. In addition, the VGAs offer very low output-referred noise performance suitable for interfacing with 10-bit ADCs.

The MAX2035 VGA is optimized for less than $\pm 0.5\text{dB}$ absolute gain error to ensure minimal channel-to-channel ultrasound beamforming focus error. The device's differential outputs are designed to directly drive ultrasound ADCs through an external passive anti-aliasing filter. A switchable clamp is also provided at each amplifier's outputs to limit the output signals, thereby preventing ADC overdrive or saturation.

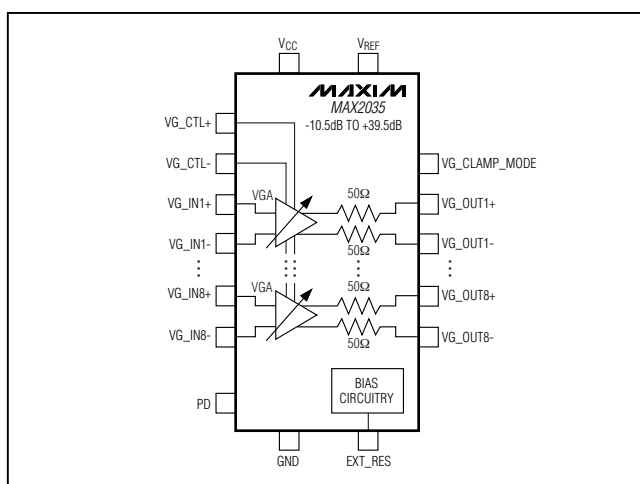
Dynamic performance of the device is optimized to reduce distortion to support second-harmonic imaging. The device achieves a second-harmonic distortion specification of -62dBc at $V_{\text{OUT}} = 1.5\text{V}_{\text{P-P}}$ and $f_{\text{IN}} = 5\text{MHz}$, and an ultrasound-specific* two-tone third-order intermodulation distortion specification of -52dBc at $V_{\text{OUT}} = 1.5\text{V}_{\text{P-P}}$ and $f_{\text{IN}} = 5\text{MHz}$.

The MAX2035 operates from a $+5.0\text{V}$ power supply, consuming only $127\text{mW}/\text{channel}$. The device is available in a 100-pin TQFP package with an exposed paddle. Electrical performance is guaranteed over a 0°C to $+70^\circ\text{C}$ temperature range.

Applications

Ultrasound Imaging Sonar

Functional Diagram



Features

- ◆ **8-Channel Configuration**
- ◆ **High Integration for Ultrasound Imaging Applications**
- ◆ **Pin Compatible with the MAX2036 Ultrasound VGA Plus CW Doppler Beamformer**
- ◆ **Maximum Gain, Gain Range, and Output-Referred Noise Optimized for Interfacing with 10-Bit ADCs**
 - Maximum Gain of 39.5dB
 - Total Gain Range of 50dB
 - 60nV/ $\sqrt{\text{Hz}}$ Ultra-Low Output-Referred Noise at 5MHz
 - Pin-for-Pin 12-Bit Compatibility Supported By MAX2037/MAX2038
- ◆ **$\pm 0.5\text{dB}$ Absolute Gain Error**
- ◆ **Switchable Output VGA Clamp Eliminating ADC Overdrive**
- ◆ **Fully Differential VGA Outputs for Direct ADC Drive**
- ◆ **Variable Gain Range Achieves 50dB Dynamic Range**
- ◆ **-62dBc HD2 at $V_{\text{OUT}} = 1.5\text{V}_{\text{P-P}}$ and $f_{\text{IN}} = 5\text{MHz}$**
- ◆ **Two-Tone Ultrasound-Specific* IMD3 of -52dBc at $V_{\text{OUT}} = 1.5\text{V}_{\text{P-P}}$ and $f_{\text{IN}} = 5\text{MHz}$**
- ◆ **127mW Consumption per Channel**

*See the Ultrasound-Specific IMD3 Specification in the Applications Information section.

Ordering Information

PART	TEMP RANGE	PIN-PACKAGE	PKG CODE
MAX2035CCQ-D	0°C to $+70^\circ\text{C}$	100 TQFP-EP† (14mm x 14mm)	C100E-3
MAX2035CCQ-TD	0°C to $+70^\circ\text{C}$	100 TQFP-EP† (14mm x 14mm)	C100E-3
MAX2035CCQ+D	0°C to $+70^\circ\text{C}$	100 TQFP-EP† (14mm x 14mm)	C100E-3
MAX2035CCQ+TD	0°C to $+70^\circ\text{C}$	100 TQFP-EP† (14mm x 14mm)	C100E-3

†EP = Exposed paddle.

+Denotes lead-free package.

T = Tape-and-reel package.

D = Dry packing.



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ABSOLUTE MAXIMUM RATINGS

V _{CC} , V _{REF} to GND	-0.3V to +5.5V	Operating Temperature Range.....	0°C to +70°C
Any Other Pins to GND.....	-0.3V to (V _{CC} + 0.3V)	Junction Temperature	+150°C
VGA Differential Input Voltage (V _{GIN_+} - V _{GIN_-}).....	8.0V _{P-P}	θ _{JC}	+2°C/W
Analog Gain-Control Input Differential Voltage (V _{G_CTL+} - V _{G_CTL-}).....	8.0V _{P-P}	θ _{JA}	+22°C/W
Continuous Power Dissipation (T _A = +70°C) 100-Pin TQFP		Storage Temperature Range	-40°C to +150°C
(derated 45.5mW/°C above +70°C).....	3636.4mW	Lead Temperature (soldering, 10s)	+300°C

Stresses beyond those listed under "Absolute Maximum Ratings" may cause permanent damage to the device. These are stress ratings only, and functional operation of the device at these or any other conditions beyond those indicated in the operational sections of the specifications is not implied. Exposure to absolute maximum rating conditions for extended periods may affect device reliability.

DC ELECTRICAL CHARACTERISTICS

(Figure 2, V_{CC} = V_{REF} = 4.75V to 5.25V, GND = 0V, PD = 0, no RF signals applied, capacitance to GND at each of the VGA differential outputs is 60pF, differential capacitance across the VGA outputs is 10pF, R_L = 1kΩ, T_A = 0°C to +70°C. Typical values are at V_{CC} = V_{REF} = 5V, T_A = +25°C, unless otherwise noted.) (Note 1)

PARAMETER	SYMBOL	CONDITIONS		MIN	TYP	MAX	UNITS
Supply Voltage Range	V _{CC}			4.75	5	5.25	V
V _{CC} External Reference Voltage Range	V _{REF}	(Note 2)		4.75	5	5.25	V
Total Power-Supply Current		Refers to V _{CC} supply current plus V _{REF} current	PD = 0		204	231	mA
			PD = 1		27	33	
V _{CC} Supply Current	I _{VCC}				192	216	mA
V _{REF} Current	I _{REF}				12	15	mA
Current Consumption per Amplifier Channel		Refers to V _{CC} supply current			24	27	mA
Differential Analog Control Voltage Range		Minimum gain			+2		V _{P-P}
		Maximum gain			-2		
Differential Analog Control Common-Mode Voltage	V _{CM}			2.85	3.0	3.15	V
Analog Control Input Source/Sink Current					4.5	5	mA
LOGIC INPUTS							
CMOS Input-High Voltage	V _{IH}			2.0			V
CMOS Input-Low Voltage	V _{IL}					0.8	V

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AC ELECTRICAL CHARACTERISTICS

(Figure 2, $V_{CC} = 4.75V$ to $5.25V$, $V_{CM} = 3/5 V_{CC}$, $V_{REF} = 5.0V$, $GND = 0V$, $PD = 0$, $VG_CLAMP_MODE = 1$, $f_{RF} = 5MHz$, capacitance to GND at each of the VGA differential outputs is $60pF$, differential capacitance across the VGA outputs is $10pF$, $R_L = 1k\Omega$, $T_A = 0^\circ C$ to $+70^\circ C$. Typical values are at $V_{CC} = V_{REF} = 5V$, $T_A = +25^\circ C$, unless otherwise noted.) (Note 1)

PARAMETER	SYMBOL	CONDITIONS		MIN	TYP	MAX	UNITS
Large-Signal Bandwidth	f-3dB	$V_{OUT} = 1.5V_{P-P}$, 3dB bandwidth, gain = 20dB	Differential output capacitance is $10pF$, capacitance to GND		17		MHz
			No capacitive load, $R_L = 1k\Omega$		22		
Differential Input Resistance	R_{IN}			170	200	230	Ω
Input Effective Capacitance	C_{IN}	$f_{RF} = 10MHz$, each input to ground			15		pF
Differential Output Resistance	R_{OUT}				100		Ω
Maximum Gain					39.5		dB
Minimum Gain					-10.5		dB
Gain Range					50		dB
Absolute Gain Error		$T_A = +25^\circ C$, $-2.0V < VG_CTL < -1.8V$			± 0.6		dB
		$T_A = +25^\circ C$, $-1.8V < VG_CTL < +1.2V$			± 0.5		
		$T_A = +25^\circ C$, $+1.2V < VG_CTL < +2.0V$			± 1.2		
VGA Gain Response Time		50dB gain change to within 1dB final value			1		μs
Input-Referred Noise		VG_CTL set for maximum gain, no input signal			2		nV/\sqrt{Hz}
Output-Referred Noise		VG_CTL set for +20dB of gain	No input signal		60		nV/\sqrt{Hz}
			$V_{OUT} = 1.5V_{P-P}$, 1kHz offset		120		
Second Harmonic	HD2	$VG_CLAMP_MODE = 1$, VG_CTL set for +20dB of gain, $f_{RF} = 5MHz$, $V_{OUT} = 1.5V_{P-P}$		-55	-62		dBc
		$VG_CLAMP_MODE = 1$, VG_CTL set for +20dB of gain, $f_{RF} = 10MHz$, $V_{OUT} = 1.5V_{P-P}$			-62		
Third-Order Intermodulation Distortion	IMD3	VG_CTL set for +20dB of gain, $f_{RF1} = 5MHz$, $f_{RF2} = 5.01MHz$, $V_{OUT} = 1.5V_{P-P}$ (Note 3)		-40	-52		dB

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AC ELECTRICAL CHARACTERISTICS (continued)

(Figure 2, $V_{CC} = 4.75V$ to $5.25V$, $V_{CM} = 3/5 V_{CC}$, $V_{REF} = 5.0V$, $GND = 0V$, $PD = 0$, $VG_CLAMP_MODE = 1$, $f_{RF} = 5MHz$, capacitance to GND at each of the VGA differential outputs is $60pF$, differential capacitance across the VGA outputs is $10pF$, $R_L = 1k\Omega$, $T_A = 0^\circ C$ to $+70^\circ C$. Typical values are at $V_{CC} = V_{REF} = 5V$, $T_A = +25^\circ C$, unless otherwise noted.) (Note 1)

PARAMETER	SYMBOL	CONDITIONS	MIN	TYP	MAX	UNITS
Channel-to-Channel Crosstalk		$V_{OUT} = 1V_{P-P}$ differential, $f_{RF} = 10MHz$, VG_CTL set for $+20dB$ of gain		-80		dB
Maximum Output Voltage at Clamp ON		$VG_CLAMP_MODE = 0$, VG_CTL set for $+20dB$ of gain, $350mV_{P-P}$ differential input		2.2		V_{P-P} differential
Maximum Output Voltage at Clamp OFF		$VG_CLAMP_MODE = 1$, VG_CTL set for $+20dB$ of gain, $350mV_{P-P}$ differential input		3.4		V_{P-P} differential

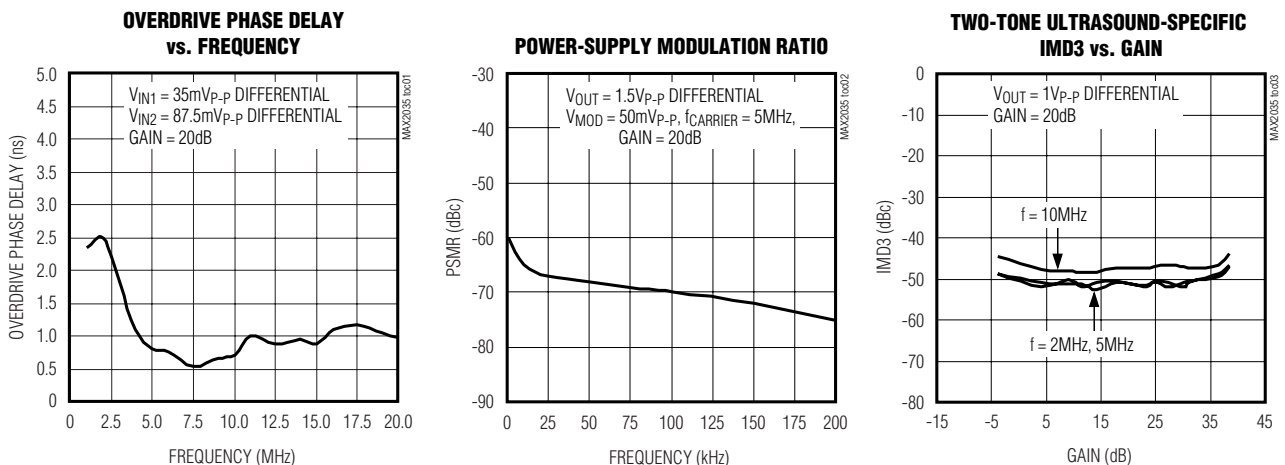
Note 1: Specifications at $T_A = +25^\circ C$ and $T_A = +70^\circ C$ are guaranteed by production test. Specifications at $T_A = 0^\circ C$ are guaranteed by design and characterization.

Note 2: Noise performance of the device is dependent on the noise contribution from the supply to V_{REF} . Use a low-noise supply for V_{REF} . V_{CC} and V_{REF} can be connected together to share the same supply voltage if the supply for V_{CC} exhibits low noise.

Note 3: See the *Ultrasound-Specific IMD3 Specification* section.

Typical Operating Characteristics

(Figure 2, $V_{CC} = V_{REF} = 4.75V$ to $5.25V$, $GND = 0V$, $PD = 0$, $VG_CLAMP_MODE = 1$, $f_{RF} = 5MHz$, capacitance to GND at each of the VGA differential outputs is $60pF$, differential capacitance across the VGA outputs is $10pF$, $R_L = 1k\Omega$, $T_A = 0^\circ C$ to $+70^\circ C$. Typical values are at $V_{CC} = V_{REF} = 5V$, $V_{CM} = 3.0V$, $T_A = +25^\circ C$, unless otherwise noted.)



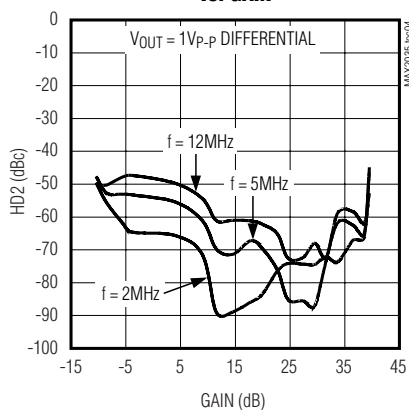
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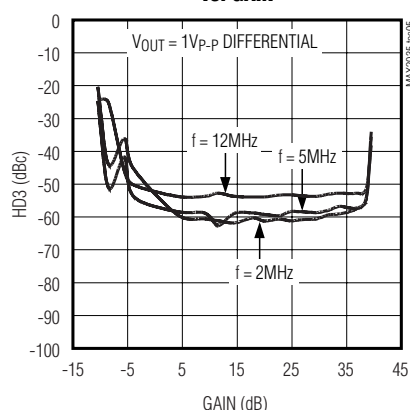
Typical Operating Characteristics (continued)

(Figure 2, $V_{CC} = V_{REF} = 4.75V$ to $5.25V$, $GND = 0V$, $PD = 0$, $VG_CLAMP_MODE = 1$, $f_{RF} = 5MHz$, capacitance to GND at each of the VGA differential outputs is $60pF$, differential capacitance across the VGA outputs is $10pF$, $R_L = 1k\Omega$, $T_A = 0^\circ C$ to $+70^\circ C$. Typical values are at $V_{CC} = V_{REF} = 5V$, $V_{CM} = 3.0V$, $T_A = +25^\circ C$, unless otherwise noted.)

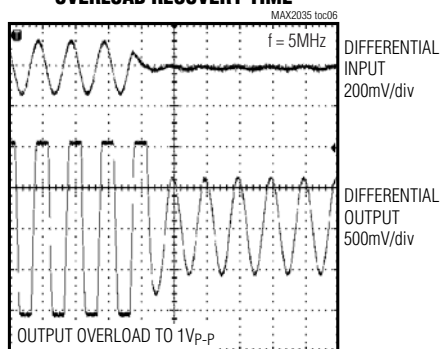
SECOND-HARMONIC DISTORTION vs. GAIN



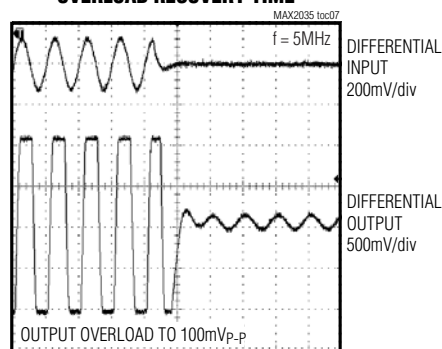
THIRD-HARMONIC DISTORTION vs. GAIN



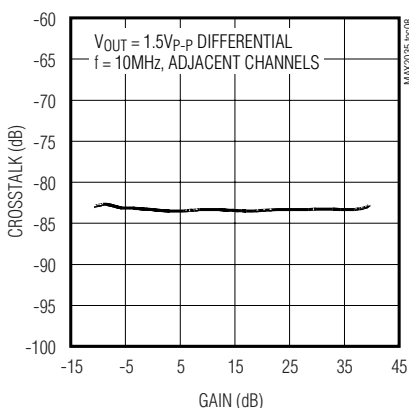
OVERLOAD RECOVERY TIME



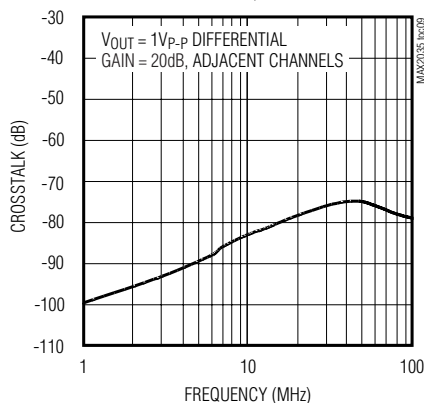
OVERLOAD RECOVERY TIME



CHANNEL-TO-CHANNEL CROSSTALK vs. GAIN



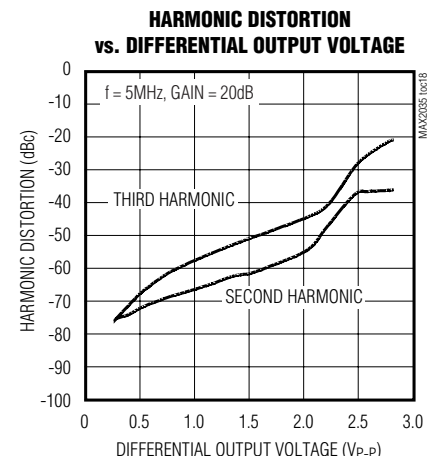
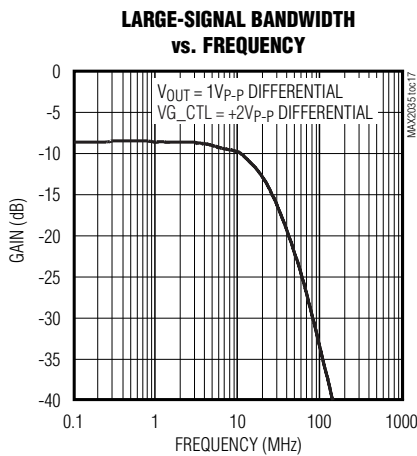
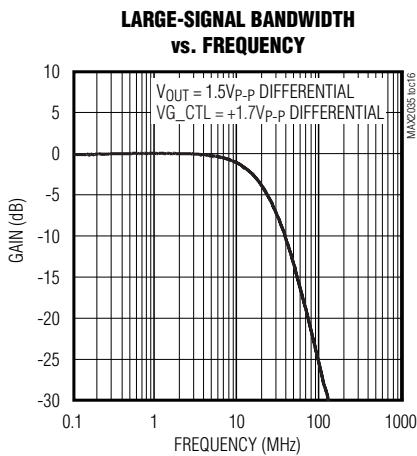
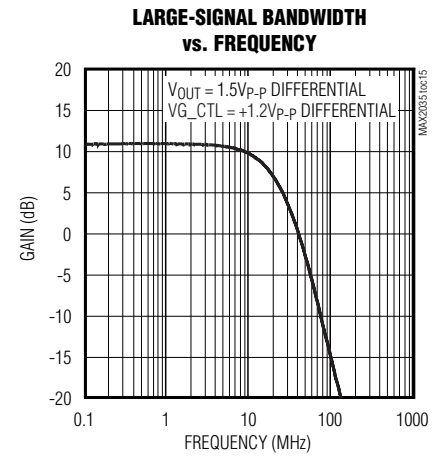
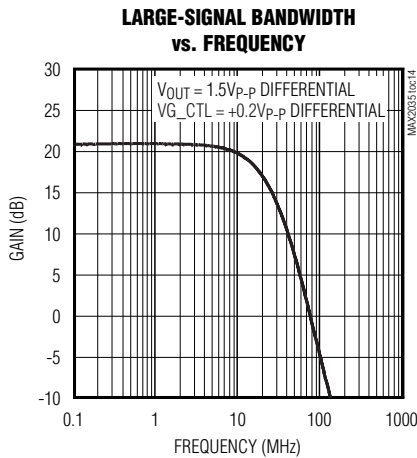
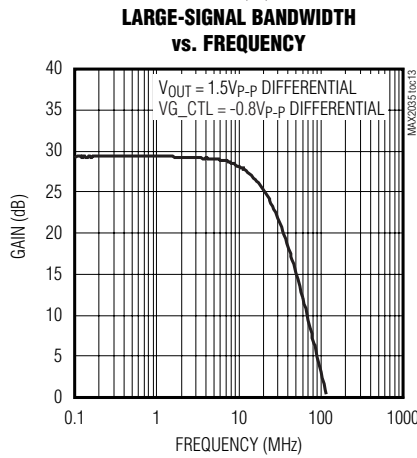
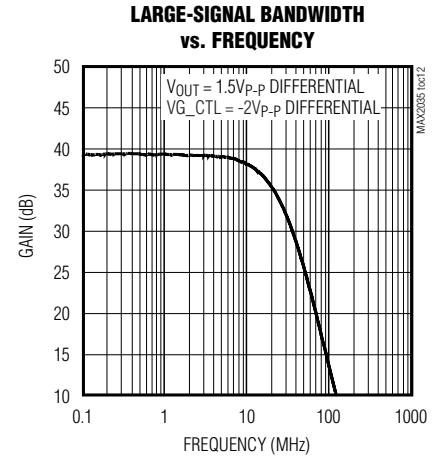
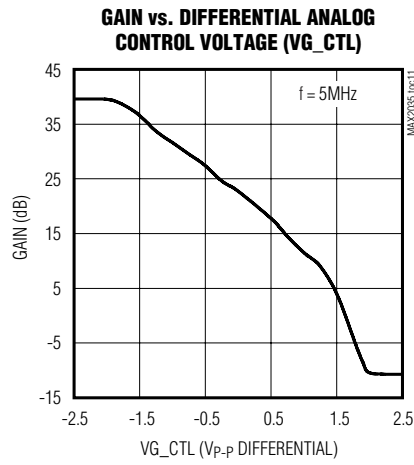
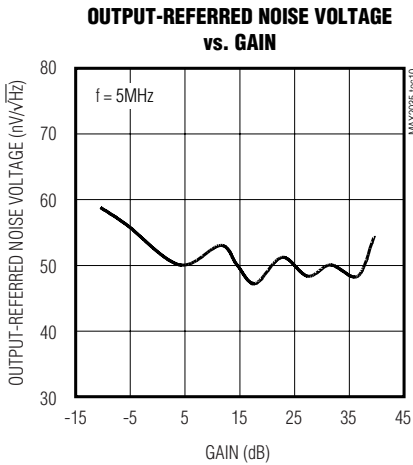
CHANNEL-TO-CHANNEL CROSSTALK vs. FREQUENCY



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Typical Operating Characteristics (continued)

(Figure 2, $V_{CC} = V_{REF} = 4.75V$ to $5.25V$, $GND = 0V$, $PD = 0$, $VG_CLAMP_MODE = 1$, $f_{RF} = 5MHz$, capacitance to GND at each of the VGA differential outputs is $60pF$, differential capacitance across the VGA outputs is $10pF$, $R_L = 1k\Omega$, $T_A = 0^\circ C$ to $+70^\circ C$. Typical values are at $V_{CC} = V_{REF} = 5V$, $V_{CM} = 3.0V$, $T_A = +25^\circ C$, unless otherwise noted.)

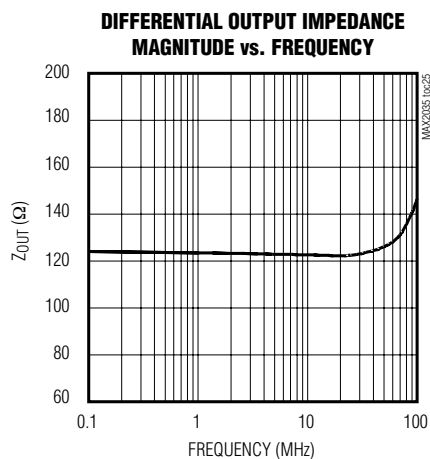
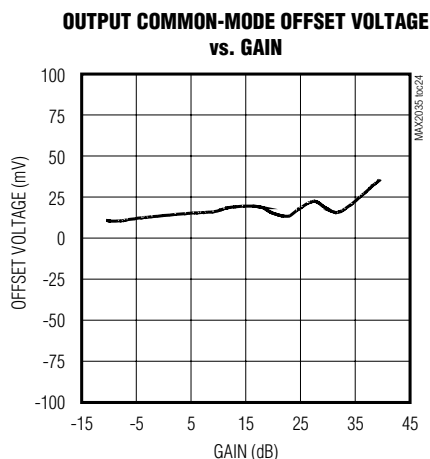
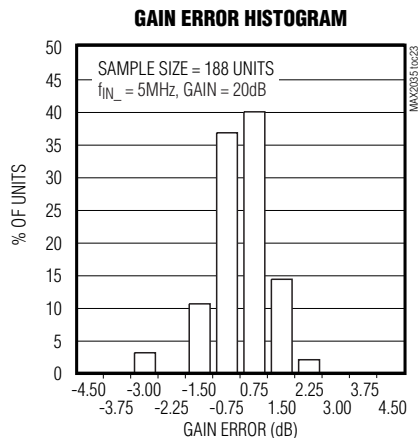
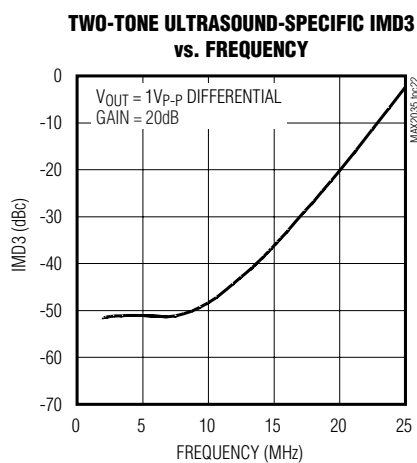
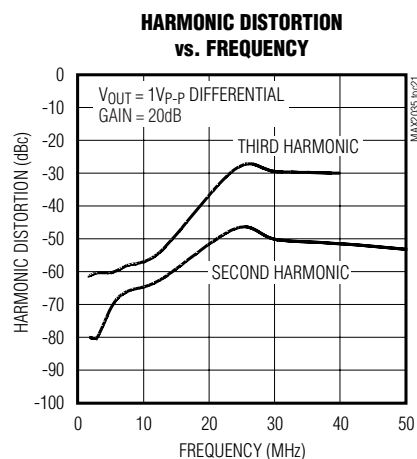
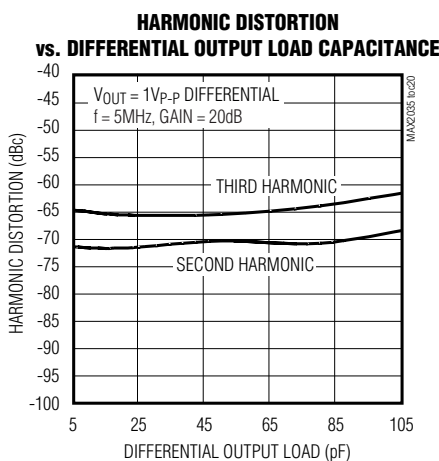
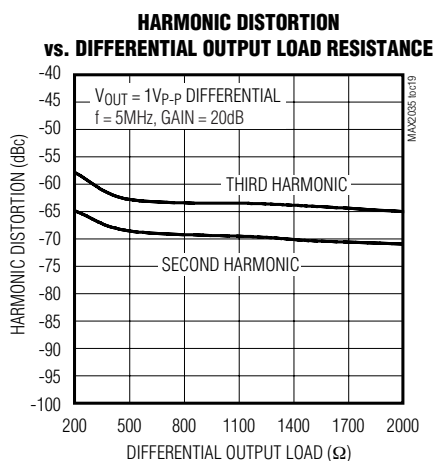


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Typical Operating Characteristics (continued)

(Figure 2, $V_{CC} = V_{REF} = 4.75V$ to $5.25V$, $GND = 0V$, $PD = 0$, $VG_CLAMP_MODE = 1$, $f_{RF} = 5MHz$, capacitance to GND at each of the VGA differential outputs is $60pF$, differential capacitance across the VGA outputs is $10pF$, $R_L = 1k\Omega$, $T_A = 0^\circ C$ to $+70^\circ C$. Typical values are at $V_{CC} = V_{REF} = 5V$, $V_{CM} = 3.0V$, $T_A = +25^\circ C$, unless otherwise noted.)



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Pin Description

PIN	NAME	FUNCTION
1, 2, 5, 6, 7, 10, 11, 12, 19, 20, 21, 24, 25, 26, 29, 30, 31, 34, 35, 36, 41, 43, 44, 45, 47, 48, 51, 55, 58, 59, 64, 65, 66, 69, 73, 76, 79, 80, 81, 83, 84, 85, 88–92, 96, 97, 98	GND	Ground
3	VGIN3-	VGA Channel 3 Inverting Differential Input
4	VGIN3+	VGA Channel 3 Noninverting Differential Input
8	VGIN4-	VGA Channel 4 Inverting Differential Input
9	VGIN4+	VGA Channel 4 Noninverting Differential Input
13	EXT_C1	External Compensation. Connect a 4.7 μ F capacitor to ground.
14	EXT_C2	External Compensation. Connect a 4.7 μ F capacitor to ground.
15	EXT_C3	External Compensation. Connect a 4.7 μ F capacitor to ground.
16, 39, 42, 46, 54, 72, 82, 87	VCC	5V Power Supply. Bypass each VCC supply to ground with 0.1 μ F capacitors as close to the pins as possible.
17	VGIN5-	VGA Channel 5 Inverting Differential Input
18	VGIN5+	VGA Channel 5 Noninverting Differential Input
22	VGIN6-	VGA Channel 6 Inverting Differential Input
23	VGIN6+	VGA Channel 6 Noninverting Differential Input
27	VGIN7-	VGA Channel 7 Inverting Differential Input
28	VGIN7+	VGA Channel 7 Noninverting Differential Input
32	VGIN8-	VGA Channel 8 Inverting Differential Input
33	VGIN8+	VGA Channel 8 Noninverting Differential Input
37, 93	VREF	5V Reference Supply. Bypass to GND with a 0.1 μ F capacitor as close to the pins as possible. Note that noise performance of the device is dependent on the noise contribution from the supply to VREF. Use a low-noise supply for VREF. VCC and VREF can be connected together to share the same supply voltage if the supply for VCC exhibits low noise.
38	EXT_RES	External Resistor. Connect a 7.5k Ω resistor to ground.
40	PD	Power-Down Switch. Drive PD high to set the device in power-down mode. Drive PD low for normal operation.
49	VGOUT8+	VGA Channel 8 Noninverting Differential Output
50	VGOUT8-	VGA Channel 8 Inverting Differential Output
52	VGOUT7+	VGA Channel 7 Noninverting Differential Output
53	VGOUT7-	VGA Channel 7 Inverting Differential Output
56	VGOUT6+	VGA Channel 6 Noninverting Differential Output
57	VGOUT6-	VGA Channel 6 Inverting Differential Output
60	VGOUT5+	VGA Channel 5 Noninverting Differential Output

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Pin Description (continued)

PIN	NAME	FUNCTION
61	VGOUT5-	VGA Channel 5 Inverting Differential Output
62	VG_CTL-	VGA Analog Gain-Control Inverting Input
63	VG_CTL+	VGA Analog Gain-Control Noninverting Input
67	VGOUT4+	VGA Channel 4 Noninverting Differential Output
68	VGOUT4-	VGA Channel 4 Inverting Differential Output
70	VGOUT3+	VGA Channel 3 Noninverting Differential Output
71	VGOUT3-	VGA Channel 3 Inverting Differential Output
74	VGOUT2+	VGA Channel 2 Noninverting Differential Output
75	VGOUT2-	VGA Channel 2 Inverting Differential Output
77	VGOUT1+	VGA Channel 1 Noninverting Differential Output
78	VGOUT1-	VGA Channel 1 Inverting Differential Output
86	VG_CLAMP_MODE	VGA Clamp Mode Enable. Drive VG_CLAMP_MODE low to enable VGA clamping. VGA output will be clamped at typically 2.2V _{P-P} differential. Drive VG_CLAMP_MODE high to disable VGA clamp mode.
94	VGIN1-	VGA Channel 1 Inverting Differential Input
95	VGIN1+	VGA Channel 1 Noninverting Differential Input
99	VGIN2-	VGA Channel 2 Inverting Differential Input
100	VGIN2+	VGA Channel 2 Noninverting Differential Input
—	EP	Exposed Paddle. Solder the exposed paddle to the ground plane using multiple vias.

Detailed Description

The MAX2035's VGAs are optimized for high linearity, high dynamic range, and low output-noise performance, making this component ideal for ultrasound-imaging applications. The VGA paths also exhibit a channel-to-channel crosstalk of -80dB at 10MHz and an absolute gain error of less than ± 0.25 dB for minimal channel-to-channel focusing error in an ultrasound system. Each VGA path includes circuitry for adjusting analog gain, an output buffer with differential output ports (VGOUT₊, VGOUT₋) for driving ADCs, and differential input ports (VGIN₊, VGIN₋) that are ideal for directly interfacing to the MAX2034 quad LNA. See the *Functional Diagram* for details.

The VGA has an adjustable gain range from -10.5dB to +39.5dB, achieving a total dynamic range of typically 50dB. The VGA gain can be adjusted with the differential gain-control input VG_CTL+ and VG_CTL-. Set the differential gain-control input voltage at -2V for maximum gain and +2V for minimum gain. The differential analog control common-mode voltage is typically 3.0V.

VGA Clamp

A clamp is provided to limit the VGA output signals to avoid overdriving the ADC or to prevent ADC saturation. Set VG_CLAMP_MODE low to clamp the VGA differential outputs at 2.2V_{P-P}. Set the VG_CLAMP_MODE high to disable the clamp.

Power Down

The device can also be powered down with PD. Set PD to logic-high for power-down mode. In power-down mode, the device draws a total supply current of 27mA. Set PD to a logic-low for normal operation.

Overload Recovery

The device is also optimized for quick overload recovery for operation under the large input signal conditions that are typically found in ultrasound input buffer imaging applications. See the *Typical Operating Characteristics* for an illustration of the rapid recovery time from a transmit-related overload.

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Applications Information

External Compensation

External compensation is required for bypassing internal biasing circuitry. Connect, as close as possible, individual $4.7\mu\text{F}$ capacitors from each pin EXT_C1, EXT_C2, and EXT_C3 (pin 13, 14, 15) to ground.

External Bias Resistor

An external resistor at EXT_RES is required to set the bias for the internal biasing circuitry. Connect, as close as possible, a $7.5\text{k}\Omega$ resistor from EXT_RES (pin 38) to ground.

Analog Input and Output Coupling

In typical applications, the MAX2035 is being driven from a low-noise amplifier (such as the MAX2034) and is typically driving a discrete differential anti-alias filter into an ADC (such as the MAX1434 octal ADC). The differential input impedance of the MAX2035 is typically 200Ω . The differential outputs are capable of driving a differential load resistance of $1\text{k}\Omega$. The output impedance is 100Ω differential. The differential outputs have a common-mode bias of approximately 3V. AC-couple these differential outputs if the next stage has a different common-mode input range.

Ultrasound-Specific IMD3 Specification

Unlike typical communications specs, the two input tones are not equal in magnitude for the ultrasound-specific IMD3 two-tone specification. In this measurement, f_1 represents reflections from tissue and f_2 represents reflections from blood. The latter reflections are typically 25dB lower in magnitude, and hence the measurement is defined with one input tone 25dB lower than the other. The IMD3 product of interest ($f_1 - (f_2 - f_1)$) presents itself as an undesired Doppler error signal in ultrasound applications. See Figure 1.

PCB Layout

The pin configuration of the MAX2035 is optimized to facilitate a very compact physical layout of the device and its associated discrete components. A typical application for this device might incorporate several devices in close proximity to handle multiple channels of signal processing.

The exposed paddle (EP) of the MAX2035's TQFP-EP package provides a low thermal-resistance path to the die. It is important that the PC board (PCB) on which the MAX2035 is mounted be designed to conduct heat from the EP. In addition, provide the EP with a low-inductance path to electrical ground. The EP **MUST** be soldered to a ground plane on the PCB, either directly or through an array of plated via holes.

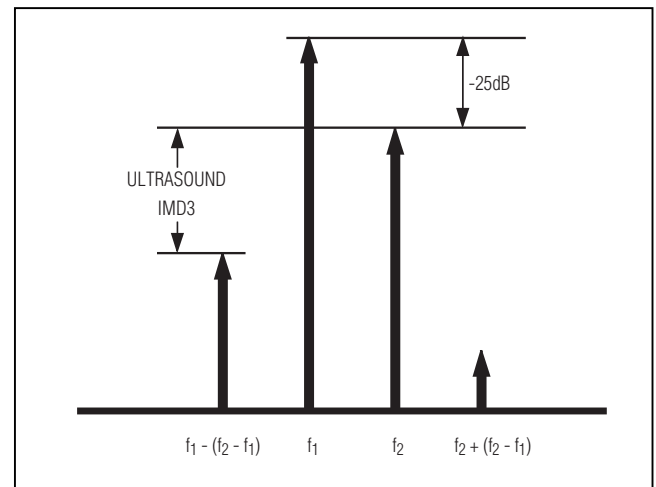


Figure 1. Ultrasound IMD3 Measurement Technique

Ultrasound Variable-Gain Amplifier

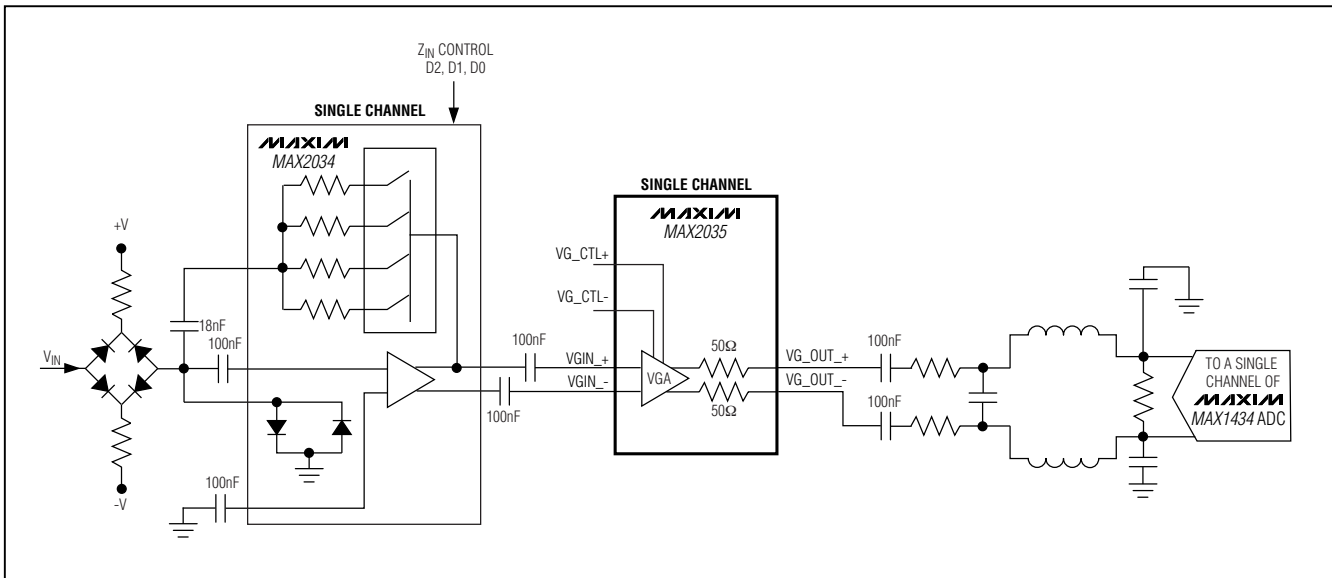
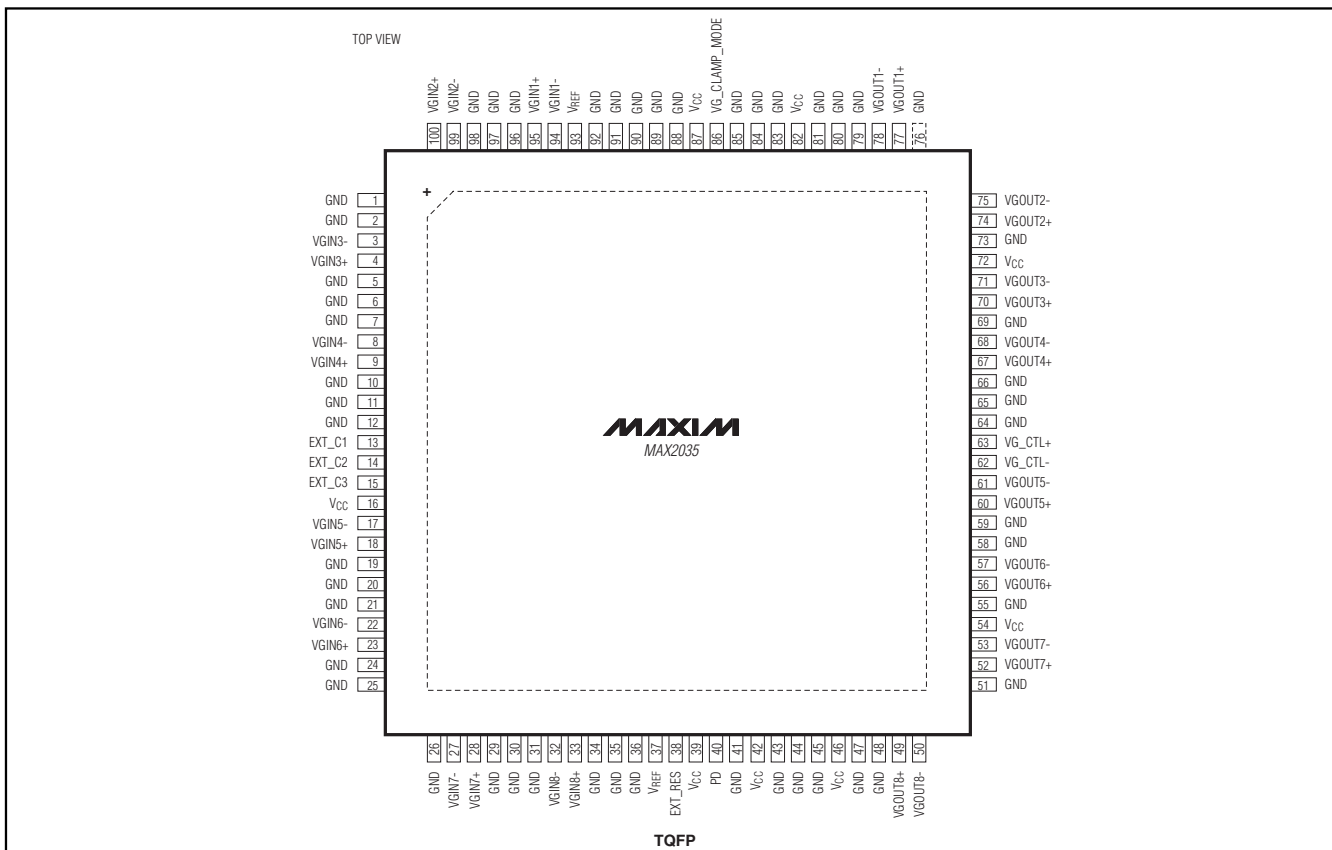


Figure 2. Typical per-Channel Ultrasound-Imaging Application

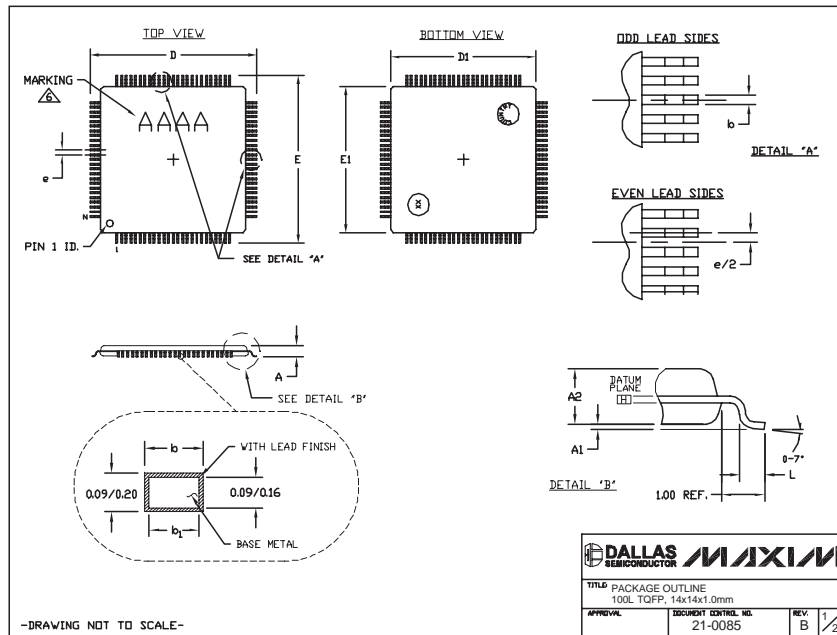
Pin Configuration



Ultrasound Variable-Gain Amplifier

Package Information

(The package drawing(s) in this data sheet may not reflect the most current specifications. For the latest package outline information go to www.maxim-ic.com/packages.)



DALLAS SEMICONDUCTOR **MAXIM**

TITLE PACKAGE OUTLINE
100L TOFP, 14x14x1.0mm

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NOTES:

- ALL DIMENSIONING AND TOLERANCING CONFORM TO ANSI Y14.5-1982.
- DATUM PLANE (A) LOCATED AT MOLD PARTING LINE AND COINCIDENT WITH LEAD, WHERE LEAD EXITS PLASTIC BODY AT BOTTOM OF PARTING LINE.
- DIMENSIONS D1 AND E1 DO NOT INCLUDE MOLD PROTRUSION. ALLOWABLE MOLD PROTRUSION IS 0.254 MM ON D1 AND E1 DIMENSIONS.
- THE TOP OF PACKAGE IS SMALLER THAN THE BOTTOM OF PACKAGE BY 0.15 MILLIMETERS.
- DIMENSION b DOES NOT INCLUDE DAMBAR PROTRUSION. ALLOWABLE DAMBAR PROTRUSION SHALL BE 0.08mm TOTAL IN EXCESS OF THE b DIMENSION AT MAXIMUM MATERIAL CONDITION.
- MARKING IS FOR PACKAGE ORIENTATION REFERENCE ONLY.
- THIS OUTLINE CONFORMS TO JEDEC PUBLICATION 95 REGISTRATION MS-026.
- LEADS SHALL BE COPLANAR WITHIN 0.08mm.

SYMBOL	TOFP PACKAGE VARIATION	
	ALL DIMENSIONS IN MILLIMETERS	
	100L, 14x14x1.0	
	MIN.	MAX.
A	1.20	1.20
A1	0.05	0.15
A2	0.95	1.05
B	15.80	16.20
D1	13.80	14.20
E	15.80	16.20
E1	13.80	14.20
L	0.45	0.75
N	100	
e	0.50 BSC.	
b	0.17	0.27
bl	0.17	0.23

-DRAWING NOT TO SCALE-

DALLAS SEMICONDUCTOR **MAXIM**

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Maxim cannot assume responsibility for use of any circuitry other than circuitry entirely embodied in a Maxim product. No circuit patent licenses are implied. Maxim reserves the right to change the circuitry and specifications without notice at any time.

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